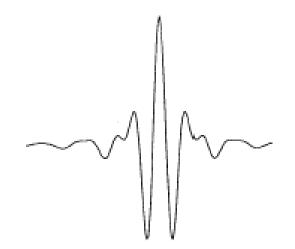
ME 252 B

Computational Fluid Dynamics: Wavelet transforms and their applications to turbulence

Marie Farge¹ & Kai Schneider²



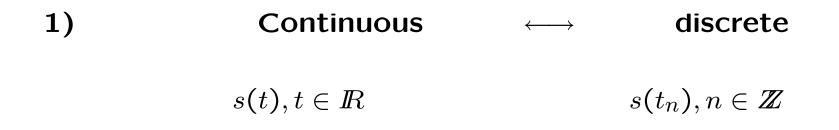
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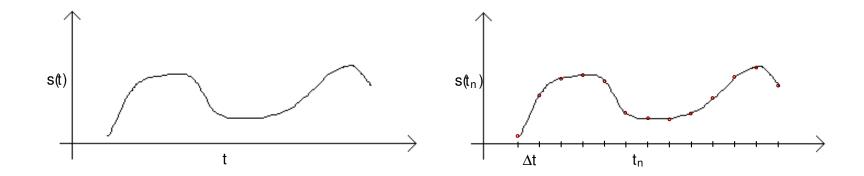
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Classification of signals (1d)

$$s:t
ightarrow s(t)$$
 with $t\in I\!\!R,\ s(t)\in I\!\!R$ or ${
m I}\!\!C$

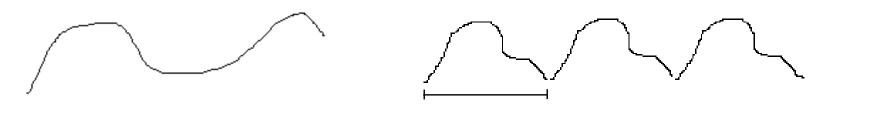




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2) Nonperiodic \leftrightarrow periodic

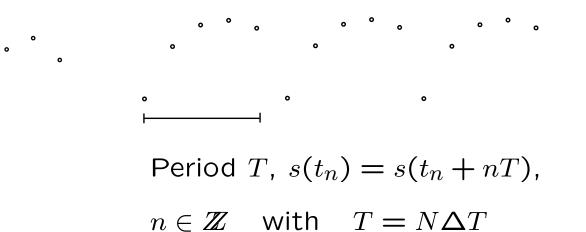
• continuous



Period T, $s(t) = s(t + nT), n \in \mathbb{Z}$

• discrete

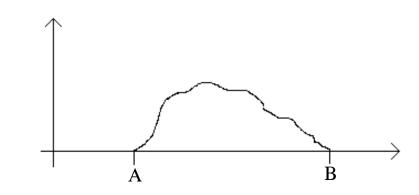
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3) Compact support

 $s(t) \neq 0$ for $t \in [A, B]$ and s(t) = 0 else



- 4) Signals with finite energy
 - continuous, nonperiodic

$$E = \int_{-\infty}^{\infty} |s(t)|^2 dt < \infty$$

• continuous, periodic

$$E = \int_0^T |s(t)|^2 dt < \infty$$

• discrete, nonperiodic

$$E = \sum_{n = -\infty}^{\infty} |s(t_n)|^2 < \infty$$

• discrete, periodic

$$E = \sum_{n=0}^{N-1} |s(t_n)|^2 < \infty$$

For mathematicians: spaces of square-integrable functions (norm + scalar product)

$$s(t) \in L^2(\mathbb{R}), L^2(\mathbb{T}), l^2(\mathbb{R}), l^2(\mathbb{T})$$
 where $\mathbb{T} = \mathbb{R}/\mathbb{Z}$

5) Absolutely integrable signals

$$S = \int_{-\infty}^{\infty} |s(t)| dt < \infty \quad s(t) \in L^{1}(\mathbb{R})$$

Classification of signals (higher dimensions)

2d
$$\longrightarrow$$
 images $s(\vec{x}) = s(x, y), x, y \in \mathbb{R}$ or $s(m, n), n, m \in \mathbb{Z}$

3d, n-d

scalar-valued \longleftrightarrow vector-valued signals

- temperature velocity
- pressure RGB signal

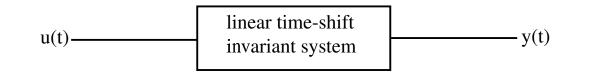
 \implies similar classification possible.

The Fourier transform Motivation

- representation of signals with sine and cosine functions
- transformation of signals into the frequency plane
- fast algorithms (FFT), $N \log_2 N$ complexity
- correlation and convolution can be efficiently computed in the frequency domain
- system theory:

sine and cosine are eigenfunctions of linear time-shift invariant systems

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$$u(t) = \sin 2\pi ft \qquad \qquad y(t) = a \sin(2\pi ft + \phi)$$

$$\cos 2\pi ft \qquad \qquad a \cos(2\pi ft + \psi)$$

For simplification one uses complex exponentials:

$$e^{it} = \cos t + i \sin t$$

Recall complex numbers: $z \in \mathbb{C}, z = x + iy = re^{i\theta}$
 $x = \Re z, y = \Im z$
 $r^2 = x^2 + y^2, \theta = \arctan y/x$

Recall trigonometric polynomials:

$$s(t) = \sum_{k \ge 0} a_k \cos 2\pi kt + b_k \sin 2\pi kt$$

Fourier transforms

1) Continuous signals

We consider an absolutely integrable signal $s(t) \in L^1(\mathbb{R})(\cap L^2(\mathbb{R}))$, $t, s \in \mathbb{R}$

The Fourier transform is defined as:

$$\widehat{S}(f) = \int_{-\infty}^{\infty} s(t)e^{-i2\pi ft}dt$$
$$= \int_{-\infty}^{\infty} s(t)\cos 2\pi ft dt + i\int_{-\infty}^{\infty} s(t)\sin 2\pi ft dt$$

Note that in general $\widehat{S}(f) \in \mathbb{C}$.

Define modulus $|\hat{S}(f)|$ and phase $\phi = \arctan \Im \hat{S}(f) / \Re \hat{S}(f)$

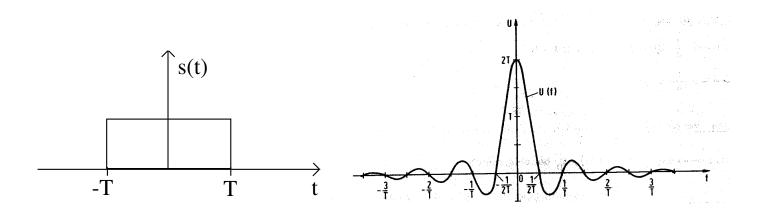
The inverse Fourier transform is defined as:

$$s(t) = \int_{-\infty}^{\infty} \widehat{S}(f) e^{i2\pi ft} df$$

Example:

$$s(t) = \begin{cases} 1 & \text{for } -T \le t \le T, \\ 0 & \text{elsewhere} \end{cases}$$
(1)

$$\widehat{S}(f) = \frac{\sin 2\pi fT}{\pi f}$$



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2) Properties

a) scaling

$$s(at) \iff \frac{1}{|a|}\widehat{S}(\frac{f}{a}) \quad a \in \mathbb{R}, a \neq 0$$
$$\widehat{S}(af) \iff \frac{1}{|a|}s(\frac{t}{a})$$

b) time-shift

$$s(t-t_0) \iff \exp(-i2\pi f t_0) \widehat{S}(f) \quad t_0 \in \mathbb{R}$$

c) frequency-shift

 $\widehat{S}(f-f_0) \iff \exp(i2\pi f_0 t)s(t) \quad f_0 \in \mathbb{R}$

d) differentiation (with respect to time) If s(t) is n-times continuously differentiable and $s^{(n)}(t) \in L^1(\mathbb{R})$, then

$$s^{(n)}(t) \iff (i2\pi f)^n \widehat{S}(f)$$

e) differentiation (with respect to frequency) If $t^m s(t) \in L^1(\mathbb{R})$ for m = 0, 1, ..., M, then $\widehat{S}^{(m)}(f)$ exists and $(-i2\pi t)^m s(t) \iff \widehat{S}^{(m)}(f)$

f) multiple application of the Fourier transform

$$\mathcal{F}\{s(t)\}(f) = \int_{-\infty}^{\infty} s(t)e^{-i2\pi ft}dt = \widehat{S}(f)$$

$$\mathcal{F}^{2}\{\{s(t)\}(f)\}(t) = \mathcal{F}\{\widehat{S}(f)\}(t) = \int_{-\infty}^{\infty} \widehat{S}(f)e^{-i2\pi ft}df = s(-t)$$

 $\longrightarrow \mathcal{F}^2$ corresponds to time inversion and hence $\mathcal{F}^4 =$ Identity

 $\mathcal{F}^3 = \mathcal{F}^{-1} = \mathcal{F}^*$ (inverse Fourier transform) Remark: The Fourier transform is a cyclic operator of 4th degree.

g) convolution given $s_1(t)$ and $s_2(t)$ with $s_1(t) \in L^2(\mathbb{R})$ and $s_2(t) \in L^{\infty}(\mathbb{R})$. $s_1(t) \star s_2(t) = \int_{-\infty}^{\infty} s_1(\tau) s_2(t-\tau) d\tau$

* commutes, i.e. $s_1 \star s_2 = s_2 \star s_1$ * is associative, i.e. $s_1 \star s_2 \star s_3 = s_1 \star (s_2 \star s_3) = (s_1 \star s_2) \star s_3$

$$s_1(t) \iff \widehat{S}_1(f) \text{ and } s_2(t) \iff \widehat{S}_2(f)$$

 $s_1(t) \star s_2(t) \iff \widehat{S}_1(f)\widehat{S}_2(f)$

h) correlation

- cross-correlation: $s_1(t), s_2(t) \in L^2(\mathbb{R})$

$$\phi_{12}(t) = \int_{-\infty}^{\infty} s_1(\tau) s_2(t+\tau) d\tau = s_1(t) \star s_2(-t)$$

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$$\phi_{21}(t) = \int_{-\infty}^{\infty} s_1(t+\tau) s_2(\tau) d\tau = s_1(-t) \star s_2(t)$$

If $\widehat{S}_1(f)$ and $\widehat{S}_2(f)$ exist, then

$$\widehat{\Phi}_{12}(f) = \mathcal{F}\{\phi_{12}(t)\}(f) = \mathcal{F}\{s_1(t) \star s_2(-t)\}(f)$$
$$= \mathcal{F}\{s_1(t) \star \mathcal{F}^2\{s_2(t)\}\}(f) = \mathcal{F}\{s_1(t)\}(f)\mathcal{F}^3\{s_2(t)\}\}(f)$$
$$= \widehat{S}_1(f)\widehat{S}_2^{\star}(f)$$

and analogously

$$\widehat{\Phi}_{21}(f) = \widehat{S}_1^{\star}(f)\widehat{S}_2(f)$$

i) autocorrelation $s_1(t) \in L^2(\mathbb{R})$

$$\phi_{11}(t) = \int_{-\infty}^{\infty} s_1(\tau) s_1(t+\tau) d\tau = s_1(t) \star s_1(-t)$$

and with $s_1(t) \iff \widehat{S}_1(f)$ we obtain in frequency space $\widehat{\Phi}_{11}(f) = \mathcal{F}\{\phi_{11}(t)\}(f) = \widehat{S}_1(f)\widehat{S}_1^{\star}(f) = |\widehat{S}_1(f)|^2$

j) multiplication

$$s_1(t)s_2(t) \quad \iff \quad \widehat{S}_1(f) \star \widehat{S}_2(f) = \int_{-\infty}^{\infty} \widehat{S}_1(\xi) \widehat{S}_2(f-\xi) d\xi$$

k) Parseval's identity

$$\int_{-\infty}^{\infty} s_1(t)s_2(t)dt = \int_{-\infty}^{\infty} \hat{S}_1(f)\hat{S}_2(-f)df$$
$$\longrightarrow \int_{-\infty}^{\infty} s_1(t)s_2^{\star}(t)dt = \int_{-\infty}^{\infty} \hat{S}_1(f)\hat{S}_2^{\star}(f)df$$
and in particular for $s_1 = s_2 = s \iff \hat{S}(f)$ we have
$$\int_{-\infty}^{\infty} |s(t)|^2 dt = \int_{-\infty}^{\infty} |\hat{S}(f)|^2 df$$

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I) energy spectrum $E(f) = |\hat{S}(f)|^2$ and $E = \int_0^\infty E(f)df$ E(f) is called spectral energy density, or energy spectrum.

m) symmetries

$$s(t) = s_{even}(t) + s_{odd}(t)$$

with $s_{even}(t) = \frac{1}{2}(s(t) + s(-t))$ and $s_{odd}(t) = \frac{1}{2}(s(t) - s(-t))$ Decomposing the corresponding Fourier transform into real and imaginary part we obtain:

 $\widehat{S}(f) = \widehat{S}_{r}(f) + i\widehat{S}_{i}(f)$ where $\widehat{S}_{r}(f) = \Re \widehat{S}(f)$ and $\widehat{S}_{i}(f) = \Im \widehat{S}(f)$ $s_{even}(t) \iff \widehat{S}_{r}(f)$ $s_{odd}(t) \iff \widehat{S}_{i}(f)$ and additionally, we have that $\widehat{S}_{r}(f)$ is even (cosine-transform)

and $\hat{S}_i(f)$ is odd (sine-transform).

- n) real valued signals If s(t) is real valued, then we have $\widehat{S}(-f) = \widehat{S}^{\star}(f)$
- o) regularity If $s^n(t) \in L^1(\mathbb{R})$ then $\lim_{f \to \pm \infty} |(i2\pi f)^n \widehat{S}(f)| = 0$, i.e. $\widehat{S}(f) = O(|f|^{-n-\epsilon})$

Bandwidth of signals and Heisenberg's uncertainty principle

$$\theta^2 = \int_{-\infty}^{\infty} (t - t_0)^2 |s(t)|^2 dt$$

$$B^{2} = \int_{-\infty}^{\infty} (f - f_{0})^{2} |\widehat{S}(f)|^{2} df$$

where $\int_{-\infty}^{\infty} |s(t)|^2 dt = \int_{-\infty}^{\infty} |\hat{S}(f)|^2 df = 1$ and t_0 and f_0 are the center of gravity in the t/f plane, respectively:

$$t_0 = \int_{-\infty}^{\infty} t |s(t)|^2 dt \qquad f_0 = \int_{-\infty}^{\infty} f |\widehat{S}(f)|^2 df$$

Heisenberg's uncertainty principle yields:

$$heta B \geq rac{1}{4\pi}$$

Proof:

w.l.o.g. let
$$t_0 = f_0 = 0$$

Using Schwarz inequality

$$\left|\int_{-a}^{b} g_{1}(t)g_{2}(t)dt\right|^{2} \leq \int_{-a}^{b} |g_{1}(t)|^{2}dt \int_{-a}^{b} |g_{2}(t)|^{2}dt$$

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for
$$a, b \to \infty$$
 and with $g_1(t) = ts(t)$ and $g_2(t) = ds/dt$ we have

$$|\int_{-\infty}^{\infty} ts(t)ds/dtdt|^{2} \leq \int_{-\infty}^{\infty} |ts(t)|^{2}dt \quad \int_{-\infty}^{\infty} |ds/dt|^{2}dt$$

As $s \in L^{2}(\mathbb{R})$, $\lim_{t \to \pm \infty} |s(t)| \leq C \frac{1}{\sqrt{t}}$

$$\int_{-\infty}^{\infty} ts(t)ds/dtdt = -\frac{1}{2}$$

and

$$\int_{-\infty}^{\infty} |ds/dt|^2 dt = \int_{-\infty}^{\infty} |2\pi f \widehat{S}(f)|^2 df$$
$$\frac{1}{4} \le 4\pi^2 \int_{-\infty}^{\infty} t^2 |s(t)|^2 dt \int_{-\infty}^{\infty} f^2 |\widehat{S}(f)|^2 df$$

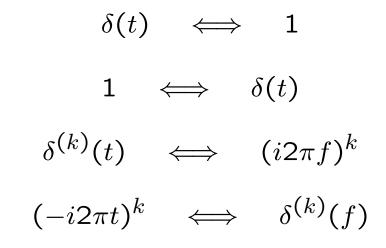
Distributions

$$\delta(t - t_0) = \begin{cases} \infty & \text{for } t = t_0, \\ 0 & \text{elsewhere} \end{cases}$$
(2)
$$\int_{-\infty}^{\infty} \delta(t - t_0) dt = 1$$

Properties:

$$\int_{-\infty}^{\infty} s(t)\delta(t-t_0)dt = s(t_0)$$
$$\sqrt{\frac{n}{\pi}}\exp(-nt^2) \longrightarrow \delta(t) \quad \text{for} \quad n \longrightarrow \infty$$
$$\sqrt{\frac{n}{\pi}}\exp(-nt^2) \iff \exp(\frac{-\pi^2 f^2}{n})$$

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Scaling:

and for k = 0

$$\delta^{(k)}(at) \iff \frac{1}{|a|} (i2\pi fa)^k$$
$$\frac{1}{|a|} \delta^{(k)}(\frac{t}{a}) \iff (i2\pi fa)^k$$
$$\frac{1}{|a|} \delta(\frac{t}{a}) \iff 1$$

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Shift:

 $\delta^{(k)}(t - t_0) \iff \exp(-i2\pi f t_0)(i2\pi f)^k$ $\exp(i2\pi f_0 t)(-i2\pi t)^k \iff \delta^{(k)}(f - f_0)$ and for k = 0 $\delta(t - t_0) \iff \exp(-i2\pi f t_0)$ $\exp(i2\pi f_0 t) \iff \delta(f - f_0)$

$$\sin(2\pi f_0 t) \iff \frac{1}{2i} (\delta(f - f_0) - \delta(f + f_0))$$
$$\cos(2\pi f_0 t) \iff \frac{1}{2} (\delta(f - f_0) + \delta(f + f_0))$$

Convolution:

 $s(t) \iff \widehat{S}(f)$ $\delta^{(k)}(t) \star s(t) \iff (i2\pi fa)^k \widehat{S}(f)$ and for k = 0

$$\delta(t) \star s(t) = s(t)$$

$$\delta(t - t_0) \star s(t) = s(t - t_0)$$

Sampling theorem: Let $s(t) \in L^1(\mathbb{R})$ with $\widehat{S}(f) = 0$ for $|f| > f_c$. Then we have

$$s(t) = \sum_{n=0}^{\infty} s(nT) \frac{\sin \pi (t - nT)/T}{\pi (t - nT)/T} \qquad \text{for} \quad T \le \frac{1}{2f_c}$$

2) Periodic signals

Periodic signal $s(t) = s(t + nT), t \in \mathbb{R}, n \in \mathbb{Z}$ with period T,

Discrete Fourier coefficients $\hat{S}_k, k \in \mathbb{Z}$ with $\hat{S}_k = \frac{1}{T} \int_0^T s(t) e^{-i2\pi kt} dt$

and $s(t) = \sum_{k \in \mathbb{Z}} \widehat{S}_k e^{i2\pi kt/T}$

3) Discrete signals

Discrete signal $s_n, n \in \mathbb{Z}$

Periodic Fourier transform

$$\widehat{S}(f) = \sum_{n \in \mathbb{Z}} s_n e^{-i2\pi nf}$$

4) Discrete periodic signals

Discrete periodic signal $s_n, 0 \le n \le N-1$ with $s_n = s_{n+mN}, m \in \mathbb{Z}$

Periodic discrete Fourier transform

$$\widehat{S}_k = \frac{1}{N}\sum_{n=0}^{N-1}s_n e^{-i2\pi kn/N}, 0\leq k\leq N-1$$
 where $\widehat{S}_k = \widehat{S}_{k+mN}, m\in Z\!\!\!Z$

5) Summary

- Continuous signal $s(t), t \in I\!\!R \iff$ continuous spectrum, $\hat{S}(f), f \in I\!\!R$ $I\!\!R$
- Periodic signal $s(t), t \in T \longrightarrow$ discrete spectrum, $\widehat{S}_k, k \in Z$
- Discrete signal $s_n, n \in \mathbb{Z} \iff$ periodic spectrum, $\widehat{S}(f), f \in \mathbb{T}$
- Discrete periodic signal $s_n, 0 \le n \le N-1$ with $s_n = s_{n+mN}, m \in \mathbb{Z}$ \longleftrightarrow periodic discrete spectrum $\hat{S}_k, 0 \le k \le N-1$ and with $\hat{S}_k = \hat{S}_{k+mN}, m \in \mathbb{Z}$

Extention to higher dimensions: tensor product ansatz

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Ad Fourier transform

Continuous signals: $s(t), t \in \mathbb{R}$

$$\widehat{S}(f) = \int_{-\infty}^{\infty} s(t)e^{-i2\pi ft}dt, f \in \mathbb{R}$$
 and $s(t) = \int_{-\infty}^{\infty} \widehat{S}(f)e^{i2\pi ft}df$

Periodic signals (continuous): $\tilde{s}(t) = \tilde{s}(t + mT), m \in \mathbb{Z}$

$$\widehat{S}_k = \frac{1}{T} \int_0^T s(t) e^{-i2\pi kt/T}$$
 and $s(t) = \sum_{k \in \mathbb{Z}} \widehat{S}_k e^{i2\pi kt/T}$

with

$$e^{i2\pi kt/T} \iff \delta(f - \frac{k}{T})$$

 $\widehat{S}(f) = \sum_{k \in \mathbb{Z}} \widehat{S}_k \delta(f - \frac{k}{T})$

Dirac pulse:

$$\sum_{k=-\infty}^{\infty} \delta(t - kT) = \frac{1}{T} \sum_{k=-\infty}^{\infty} \delta(f - k/T)$$

Periodisation

$$\tilde{s}(t) = \sum_{n=-\infty}^{\infty} s(t - nT)$$
$$\tilde{s}(t) = \tilde{s}(t + mT), m \in \mathbb{Z}$$
$$\tilde{s}(t) \iff \sum_{k \in \mathbb{Z}} \hat{S}_k \delta(f - \frac{k}{T})$$

with $\hat{S}_k = \frac{1}{T}\hat{S}(k/T)$

Discrete signals: $s_n, n \in \mathbb{Z}$

$$\widehat{S}(f) = \sum_{n \in \mathbb{Z}} s_n e^{-i2\pi nf}$$

Discrete periodic signals: s_n , n = 0, ..., N - 1

$$\widehat{S}_k = \sum_{n \in \mathbb{Z}} s_n e^{-i2\pi kn/N}$$

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Sampled signals:

$$s_{samp}(t) = T \sum_{k \in \mathbb{Z}} s(kT) \delta(t - kT)$$

$$f_{samp} = 2f_{limit} = \frac{1}{T}$$

$$s_{samp}(t) \iff \widehat{S}_{samp}(f) = \sum_{k=-\infty}^{\infty} \widehat{S}(f - kf_{samp})$$